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High Speed 17-tap Finite Impulse Response Filter based on Distributive Arithmetic Technique

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ABSTRACT: Low complexity and configurability is the key features in emerging communication applications in order to support multi-standards and operation modes. To obtain these features, efficient implementations of finite impulse response (FIR) filter with multiplier-less distributive arithmetic technique is proposed in this paper. In this technique consist of Look Up Table (LUT), shift register and accumulator. Based on this technique multipliers in FIR filter are removed. Multiplication is performed through shift and addition operations. The LUT can be subdivided into a number of LUT to reduce the size of the LUT for higher order filter. Each LUT operates on a different set of filter taps. Analysis on the performance of various filter orders with different address length are done using Xilinx synthesis tool. The proposed architecture provides less latency and less area compared with existing structure of FIR filter.

KEYWORDS: Finite Impulse Response (FIR), Look Up Table (LUT), Distributive Arithmetic Technique

I. INTRODUCTION

A digital filter is a system that performs mathematical operations on a sampled or discrete time signal to reduce or enhance certain aspects of that signal. One type of digital filter is FIR filter. It is a stable filter. It gives linear phase response. Pipelining and parallel processing technique is used in FIR filter. Pipelining operation takes place in an interleaved manner. Pipelining done by inserting latches (delay element) in the system. It increases the overall speed of the architecture but the hardware structure and system latency will increases. Hardware structures increases due to inserting pipelining latches. For M level pipelining M-1 delay elements required. Latency is the difference between the availability of first output in the sequential system and the pipeline system [1]. At every clock cycle it will operate multiple inputs and produced multiple outputs is called parallel processing. It required extra hardware. Both pipelining and parallel processing has disadvantages. For FIR filters, output is a linear convolution of weights and inputs. For an Nth-order FIR filter, the generation of each output sample takes N+1 multiply accumulate (MAC) operations. Multiplication is strongest operation because it is repeated addition. It require large portion of chip area. Power consumption is more. Memory-based structures are more regular compared with the multiply accumulate structures; and have many other advantages, e.g., greater potential for high throughput and reduced-latency implementation and are expected to have less dynamic power consumption due to less switching activities for memory-read operations compared to the conventional multipliers. Memory based structures are well-suited for many digital signal processing (DSP) algorithms, which involve multiplication with a fixed set of coefficients. For this Distributed Arithmetic architecture used in FIR filter.

Distributed arithmetic is one way to implement convolution with multiplier less unit, where the MAC operations are replaced by a series of LUT access and summations. Distributed Arithmetic is a different approach for implementing



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digital filters. The basic idea is to replace all multiplications and additions by a table and a shifter-accumulator. LUT are the kind of logic that used in SRAM based FPGAs. Basically each look table is a bunch of single bit memory cells storing individual bit values in each of the cells. Memory access time is less in SRAM, so speed of the static RAM is high. Distributed Arithmetic provides cost-effective and area-time efficient computing structures. Digital Finite Impulse Response (FIR) filters are essential building blocks in most Digital Signal Processing (DSP) systems. A large application area is telecommunication, where filters are needed in receivers and transmitters, and an increasing portion of the signal processing is done digitally [2, 3]. However, power dissipation of the digital parts can be a limiting factor, especially in portable, battery-operated devices. Scaling of the feature sizes and supply voltages naturally helps to reduce power. For a certain technology, there are still many kinds of architectural and implementation approaches available to the designer.

II. DISTRIBUTIVE ARITHMETIC TECHNIQUE

Distributed arithmetic is one way to implement convolution with multiplier less unit, where the multiplication operations are replaced by a series of LUT access and summation.

$$Y_{H} = \begin{bmatrix} 71 & 38 & 4 & 6 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 3 \\ 4 \end{bmatrix} = 183$$

Apply DA Technique

	E .					
	1	0	0	0		
	1	1	0	1		$r_1 + r_2 + r_4$
	1	1	1	1	r_1	$r_1 + r_2 + r_3 + r_4$
<i>Y</i> =	0	0	0	0	2	0
- H	0	0	0	0	3	0
	0	1	0	0	4	r ₂
	1	0	0	0	-	<i>r</i> ₁

step-1 all the input converts' binary number, Step-2 all the binary input applied to a input buffer array so,

$P_1 = 0001$	$P_2 = 0111$
$P_3 = 1010$	$P_4 = 0000$
$P_{5} = 0000$	$P_{6} = 0010$
$P_7 = 0001$	

The entire adder array input applied to MUX so, the entire adder array input

MUX (1) =
$$0001 = Y_p(0)$$



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MUX (1) add MUX (2) = $Y_P(1)$

- = 00001
- = 01110
- + 01111

Output of the $Y_P(1)$ again right shift 1-bit and adds MUX (3) so

- = 001111
- = 101000
- + 110111

$$Y_{P}(1) + MUX(3) = Y_{P}(2)$$

Output of the $Y_P(2)$ again right shift 1-bit and adds MUX (6) so

- = 000110111
- = 00100000
- + 001110111

 $\begin{array}{l} Y_{P}\left(2\right)+MUX\left(6\right)=Y_{P}\left(3\right)\\ \text{Output of the }Y_{P}\left(3\right)\text{ again right shift 1-bit and adds MUX}\left(7\right)\\ \text{so}\\ = 001110111 \end{array}$

- = 001000000
- + 010110111

 $Y_P(3) + MUX(6) = Y_P(4)$ Total output $Y_P(4) = 010110111 = 183$



Figure 1: The block diagram of DA computation



Output Sample

Figure 2: Multiplier-less Distributive Arithmetic Based FIR Filter



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III. PROPOSED METHODOLOGY

The above technique holds good only when we go for lower order filters. For higher order filters, the size of the LUT also increases exponentially with the order of the filter. For a filter with N coefficients, the LUT have 2N values. This in turn reduces the performance.



Figure 3: View Technology of 17-tap FIR Filter



Figure 4: RTL View of 17-tap FIR Filter



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Name	Value	0 ns	200 ns	400 ns	600 ns
🕨 📷 xn2[7:0]	00001111	00000000		0000	1111
🗓 clk	0				מנותה ההתרוחות
🗓 rst	1				
ᡙ cinn	0				
▶ 📑 h0[7:0]	00010101	00000000		00010	0101
🕨 📑 h1[7:0]	00010101	00000000		00010	0101
🕨 📑 h2[7:0]	00010101	00000000		00010	0101
🕨 📑 h3[7:0]	00010101	00000000		00010	0101
🕨 📑 h4[7:0]	00010101	00000000		00010	0101
🕨 📑 h5[7:0]	00010101	00000000		00010	0101
🕨 📑 h6[7:0]	00010101	00000000		00010	0101
🕨 式 h7[7:0]	00010101	00000000		00010	0101
🕨 📑 h8[7:0]	00010101	00000000		00010	0101
🕨 📑 h9[7:0]	00010101	00000000		00010	0101
🕨 📑 h10[7:0]	00010101	00000000		00010	0101
🕨 📑 h11[7:0]	00010101	00000000		00010	0101
🕨 📑 h12[7:0]	00010101	00000000		00010	0101
h13[7:0]	00010101	00000000		00010101	
h14[7:0]	00010101	00000000		00010101	
🕨 🐻 h15[7:0]	00010101	(00000000 X		00010101	
▶ =6 h16[7:0]	00010101	(00000000 X		00010101	
yn2[15:0]	00000000011	(0000000)		00000	00001101011
lia cout	0				
Clk_period	10000 ps			10000 ps	

Figure 5: Output Waveform of 17-tap FIR Filter

Device utilization summary:

Selected Device : 7vh290thcg1155-2

Slice Logic Utilization:					
Number of Slice Registers:	188	out	of	437600	0%
Number of Slice LUTs:	1843	out	of	218800	0%
Number used as Logic:	1843	out	of	218800	0%
Slice Logic Distribution:					
Number of LUT Flip Flop pairs used:	1942				
Number with an unused Flip Flop:	1754	out	of	1942	90%
Number with an unused LUT:	99	out	of	1942	5%
Number of fully used LUT-FF pairs:	89	out	of	1942	4%
Number of unique control sets:	1				
IO Utilization:					
Number of IOs:	164				
Number of bonded IOBs:	164	out	of	300	54%
Specific Feature Utilization:					
Number of BUFG/BUFGCTRLs:	1	out	of	32	3%
			_		

Figure 6: Device Utilization Summary of 17-tap FIR Filter

Timing Summary: -----Speed Grade: -2

> Minimum period: 0.703ns (Maximum Frequency: 1421.666MHz) Minimum input arrival time before clock: 1.183ns Maximum output required time after clock: 14.882ns Maximum combinational path delay: 14.659ns

Figure 7: Timing Summary of 17-tap FIR Filter



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IV. CONCLUSION

Finite Impulse Response filter plays an important role in many Digital Signal Processing applications. In this method, the multiplier less FIR filter is implemented using Distributed Arithmetic which consists of Look Up Table and then partitioning is involved. Memory access time is less than multiplication time. LUT partition reduces memory requirements. This technique reduces the delay, area, power consumption. The performance can be further improved by pipelining all the partial tables. This architecture provides an efficient area-time power implementation which involves significantly less latency and less area-delay complexity when compared with existing structures for FIR Filter.

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